Control Design under Actuator Saturation and Multi-Rate Sampling

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Abstract

The problem of designing a stabilizing feedback controller in the presence of saturating actuators and multi-rate (asynchronous) aperiodic state measurements is studied. Specifically, we consider a scenario in which measurements of the plant states are collected at the controller end in a sporadic and asynchronous fashion. A hybrid controller is used to perform a fusion of measurements sampled at different times. In between sampling events, the controller behaves as a copy of the plant and provides a feedback control signal based on the reconstruction of the plant state. The presence of saturation at the plant input limits the value of the components of this signal to a bounded range. When a new measurement is available, the controller state undergoes an instantaneous jump. The resulting system is augmented with a set of timers triggering the arrival of new measurements and analyzed in a hybrid systems framework. Relying on Lyapunov tools for hybrid systems and techniques for control design under saturation, we propose sufficient conditions in the form of matrix inequalities to ensure regional exponential stability of a closed-set containing the origin of the plant, i.e., exponential stability with a guaranteed region of attraction. Specifically, explicit estimates of the basin of attraction are provided in the form of ellipsoidal sets. Leveraging those conditions, a design procedure based on semidefinite programming is proposed to design a stabilizing controller with maximized size of the basin attraction. The effectiveness of the proposed methodology is shown in an example.

Key words: Hybrid Systems, Saturated Feedback, Networked Control Systems, Lyapunov Stability, LMIs.

1 Introduction

Control of sampled-data systems has been attracting the attention of researchers for a long time and its interest is still vivid in the community as pointed out by the survey [21]. Indeed, capturing the effect of sampled-data information in control loops is fundamental in addressing the challenges posed by modern engineered systems, in which the use of embedded systems and communica-

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tion networks is central; see, e.g., [20] and the references therein. One of the most challenging aspects in the literature of sampled-data systems pertains to the study of aperiodic sampled-data systems, that is systems in which the exchange of information from the plant to the controller happens in a sampled-data aperiodic fashion. In fact, aperiodic sampling may be used as a modeling paradigm to capture different phenomena occurring in data networks such as dropouts, sampling jitter, etc; see [21] for a survey on sampled-data systems. Essentially three main methodologies can be found in the literature for the analysis of sampled-data systems. The first consists of a time-varying discrete-time representation of the sampled-data system. This approach is pursued, e.g., in [32,34]. The second category of methodologies builds upon a time-delay model of the sampling mechanism and the use of Lyapunov-Krasovskii functionals for stability analysis. Following this approach, sufficient conditions for closed-loop asymptotic stability of linear aperiodic sampled-data systems are proposed in [14,16]. The third group of methodologies originates from the use

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of hybrid/impulsive models. This line of research is explored in [30], where Lyapunov functions with discontinuity at the sampling times are exploited to derive sufficient conditions for stability of linear aperiodic sampled-data systems. Computationally tractable conditions for stability analysis of linear aperiodic impulsive systems are given in [4]. More general stability results for nonlinear sampled-data systems based on the hybrid systems framework in [18] are given in [31].

A common feature in the above references consists of assuming synchronous sampling of different variables. However, in certain applications, due to technological constraints, some variables may happen to be measured at different sampling rates than others, making the assumption of synchronous sampling restrictive; see, e.g., [2,29]. Because of its practical relevance, this problem has been attracting the attention of researchers for a long time [23,7] and [24,30,27] for more recent contributions. A common characteristic in all the works cited so far consists of adopting zero-order-holder devices to generate intersample control inputs, i.e., the control input is held constant between sampling events. More general "holding functions" are proposed in [17] for multi-rate discrete-time systems. In particular, in [1] the authors show how exponentially decaying holding functions can be used to tolerate larger sampling rates in linear singlerate sampled-data systems.

Another limitation omnipresent in control systems pertains to actuator saturation. Due to its practical relevance, the interplay of sampling and saturation has attracted the interest of researchers over the last decade. Synchronous sampled-data \mathcal{H}_{∞} control design for linear plants subject to input saturation is addressed in [8] via hybrid systems tools. Stability analysis and control design for linear systems under aperiodic saturated state/output feedback are tackled in [15], and later refined in [19,35], where the looped-functional approach in [5,34] is adopted.

Although the analysis of saturation in (single rate) sampled-data systems has received a lot of attention, despite the relevance of asynchronous sample-data systems in practical applications, to the best of our knowledge, no results are found in the literature to deal with actuator saturation in an asynchronous sampling setting. The objective of this paper is to fill this gap. To this end, we propose a controller that achieves local exponential stability by using information sampled in a sporadic multi-rate (asynchronous) fashion. Our main contribution consists of sufficient conditions for the design of such a controller in a systematic way, i.e., via the solution to suitable optimization problems. As opposed to the case of networked control systems [6] and event-triggered control systems [13], we assume that the communication between plant and controller is not ruled by a "stabilizing" protocol and that only some upper bounds on the largest sampling rates are available.

This is typically the case when variables are measured via sensors working at different sampling-rates or in the presence of network unreliabilities in transmission links. Our approach in solving this problem can be summarized as follows:

- We propose a hybrid controller that merges measurements sampled at different times and that provides a feedback control signal based on the reconstruction of the plant state. In between sampling events, the controller behaves as a copy of the plant, while when a new measurement is available, the controller state undergoes an instantaneous jump. This extends prior work in [28] to the case of input saturation.
- Because of input saturation, achieving global exponential stability is in general impossible. To overcome this problem, we focus on regional exponential stability. In particular, as shown in Section 3, by relying on the use of Lyapunov theory for hybrid systems in the framework [18], quadratic-like clock-dependent Lyapunov functions, and a characterization of the saturation based on regional sector conditions, we provide sufficient conditions in the form of matrix inequalities to assure closed-loop regional exponential stability. In this direction, one of the main contributions of this paper consists of the use of a clock-dependent Lyapunov function in a regional stabilization setting. The main difficulty to tackle in this case is that estimates of the basin of attraction (given in terms of sublevel set of the Lyapunov function) inherently depend on the "artificial" clock variable. To deal with this aspect, we adapt the local sector bound approach in [36].
- Building upon the results in Section 3, Section 4 proposes conditions that are specifically suited to controller design. One of the main contributions of this latter section is the formulation of a numerically tractable optimization problem for the design of a controller inducing basin of attraction maximization.

It is worth to mention that the use of hybrid systems tools in the setup considered in this paper makes it possible to consider the effect of controller initial conditions on the closed-loop system. This is a relevant aspect in the case of saturating actuators since a mismatch in controller and plant initial conditions may have a dramatic impact on the behavior of the closed-loop system. In most of the results proposed in the literature, controller and plant initial conditions are assumed to be identical; see [15,34] and the references therein, no assumption of this type is considered in our work. This paper extends the results in our preliminary conference paper [11]. In particular, while in [11] we only focus on stability analysis, this work pertains to control design with basin of attraction maximization. In addition, here we provide full proofs of all the results; no proofs are included in the conference paper [11]. Due to space constraints, the proof of some auxiliary results as well as preliminaries on hybrid systems are omitted and can be found in the extended version of this paper [12].

Notation

The set \mathbb{N} is the set of positive integers including zero, the set $\mathbb{N}_{>0}$ is the set of strictly positive integers, $\mathbb{R}_{>0}$ represents the set of nonnegative real numbers, \mathbb{R}^n is the Euclidean space of dimension n, $\mathbb{R}^{n \times m}$ represents the set of the $n \times m$ real matrices, $\mathbb{R}^{n}_{>0}$ is set of vectors in \mathbb{R}^{n} with positive entries, and \mathbb{S}^{n}_{+} is the set of $n \times n$ symmetric positive definite matrices, and \mathcal{D}_{\perp}^{n} denotes the set of $n \times n$ diagonal positive definite matrices. For a vector $x \in \mathbb{R}^n$, |x| denotes the Euclidean norm and x_i stands for the *i*-th component of x. Given vectors x, y, we use the equivalent notation $(x,y) = [x^{\mathsf{T}} \ y^{\mathsf{T}}]^{\mathsf{T}}$. The identity matrix of size n is denoted by \mathbf{I}_n . The symbol $\mathbf{1}_n$ denotes the all-ones vector in \mathbb{R}^n . For a matrix $A \in \mathbb{R}^{n \times m}$, A^T denotes the transpose of A, $A_{(i)}$ the *i*-th row of A, $A_{(i)}^{\mathsf{T}}$ the *i*-th row of A^{T} , and, when n = m, $\operatorname{He}(A) = A + A^{\mathsf{T}}$. For a symmetric matrix A, positive definiteness (negative definiteness) and positive semidefiniteness (negative semidefiniteness) are denoted, respectively, by $A \succ 0$ $(A \prec 0)$ and $A \succ 0$ $(A \prec 0)$. In partitioned symmetric matrices, the symbol \star stands for symmetric blocks. Given matrices A and B, $A \oplus B$ stands for the direct sum of A and B, i.e., the block-diagonal matrix hav-

ing A and B as diagonal blocks and $\bigoplus_{i=1}^n A_i$ is the direct sum of the matrices A_1, A_2, \ldots, A_n . Given $x \in \mathbb{R}^n$ and a nonempty set A, the distance of x to A is defined as $|x|_A = \inf_{y \in A} |x-y|$. For any function $z \colon \mathbb{R} \to \mathbb{R}^n$, for any $t \in \mathbb{R}$, we denote $z(t^+) \coloneqq \lim_{s \to t^+} z(s)$. Given $P \in \mathbb{S}_+^n$ and c > 0, $\mathcal{E}(P,c) \coloneqq \{x \in \mathbb{R}^n \colon x^T Px \le c\}$. Given a set S, we denote $\cos S$ the convex-hull of S. The symbol \times denotes the Cartesian product of two sets, while given sets S_1, S_2, \ldots, S_n , we use the notation $X_{i=1}^n S_i = S_1 \times S_2 \times \cdots \times S_n$. The symbol $\operatorname{rge} f$ stands for the image of the function f. Let f in f

2 Problem Formulation

2.1 System description

Consider the following continuous-time plant:

$$\begin{cases} \dot{x}_p(t) = Ax_p(t) + Bu(t) \\ u(t) = \mathfrak{sat}_{\bar{u}}(v(t)), \end{cases}$$
 (1)

where $x_p \in \mathbb{R}^n$ is the plant state, $u \in \mathbb{R}^m$ is the plant input and $v \in \mathbb{R}^m$ is the actuator input. Matrices $A \in$

 $\mathbb{R}^{n \times n}$ and $B \in \mathbb{R}^{n \times m}$ are given. The function $v \mapsto \mathfrak{sat}_{\bar{u}}(v)$ is the symmetric decentralized saturation function with saturation level $\bar{u} = (\bar{u}_1, \bar{u}_2, \dots, \bar{u}_m) \in \mathbb{R}^m_{>0}$. More precisely,

$$v \mapsto \mathfrak{sat}_{\bar{u}}(v) := (\operatorname{sat}_1(v_1), \operatorname{sat}_2(v_2), \dots, \operatorname{sat}_m(v_m)), (2)$$

where for each $v_i \in \mathbb{R}$, $\operatorname{sat}_i(v_i) := \min(|v_i|, \bar{u}_i) \operatorname{sign}(v_i)$, i = 1, ..., m.

We assume that sporadic and asynchronous measurements of the components of the plant state x_p are available. More precisely, assume that the components of the plant state x_p are clustered into $q \leq n$ subvectors $x_p^{(1)} \in \mathbb{R}^{n_1}, x_p^{(2)} \in \mathbb{R}^{n_2}, \ldots, x_p^{(q)} \in \mathbb{R}^{n_q}$, i.e., $x_p = (x_p^{(1)}, x_p^{(2)}, \ldots, x_p^{(q)})$ and that measurements of $x_p^{(i)}$ are available only at some time instances $t_k^{(i)}, k \in \mathbb{N}$, not known a priori. For each $i \in \mathcal{N} \coloneqq \{1, 2, \ldots, q\}$, we define $M_i \subset \mathbb{R}_{\geq 0}$ as the set of measurement times of $x_p^{(i)}$, i.e., $M_i = \{t_k^{(i)}\}_{k=0}^{\infty}$ and $\mathcal{M} = \bigcup_{i=1}^q M_i$ as the set of all measurement times. In this setting, our goal is to design a feedback controller ensuring regional closed-loop exponential stability of the origin of the plant (1), i.e., local exponential stability with an explicit estimate of the basin of attraction.

Inspired by [28], we consider the following controller with jumps in its state:

$$\begin{cases} \dot{\eta}(t) = A\eta(t) + B\mathfrak{sat}_{\bar{u}}(v(t)) & \text{if } t \notin \mathcal{M} \\ \eta(t^{+}) = \pi(x_{p}(t), \eta(t), t) & \text{if } t \in \mathcal{M} \\ v(t) = K\eta(t) & \forall t \in \mathbb{R}_{\geq 0} \end{cases}$$
(3)

where the vector $\eta \in \mathbb{R}^n$ represents the controller state. The matrix $K \in \mathbb{R}^{m \times n}$ is a control gain to be designed and the function $\pi \colon \mathbb{R}^{2n} \times \mathcal{M} \to \mathbb{R}^n$ is defined for all $(x_p, \eta, t) \in \mathbb{R}^{2n} \times \mathcal{M}$ as

$$\pi(x_p, \eta, t) := (\pi_1(x_p^{(1)}, \eta^{(1)}, t), \pi_2(x_p^{(2)}, \eta^{(2)}, t), \dots, \pi_q(x_p^{(q)}, \eta^{(q)}, t))$$

with
$$\pi_i(x_p^{(i)}, \eta^{(i)}, t) := \begin{cases} x_p^{(i)} & \text{if } t \in M_i \\ \eta^{(i)} & \text{otherwise} \end{cases}$$
, $\forall i \in \mathcal{N}$.

The operating principle of the controller in (3) is an estimator triggered by the new measurement of $x_p^{(i)}$. Indeed, the arrival of a new measurement $x_p^{(i)}$ for some $i \in \mathcal{N}$ triggers an instantaneous jump in the controller state η via the function π . At each jump, the measured vector $x_p^{(i)}$ is instantaneously stored in $\eta^{(i)}$. If different components of x_p are measured simultaneously, then all the corresponding components of η get updated simultaneously. Then, in between consecutive measurements, η is continuously updated according to continuous-time

dynamics mimicking the plant dynamics and its value is continuously used in place of the plant state $x_p^{(i)}$ in a static state-feedback controller scheme.

The following assumption on the event times is considered.

Assumption 1 For each $i \in \mathcal{N}$, the sequence $\{t_k^{(i)}\}_{k=1}^{\infty}$ is unbounded and such that there exist two positive real numbers $T_1^{(i)} \leq T_2^{(i)}$ such that

$$0 \le t_1^{(i)} \le T_2^{(i)} T_1^{(i)} \le t_{k+1}^{(i)} - t_k^{(i)} \le T_2^{(i)} \forall k \in \mathbb{N}.$$
 (4)

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Remark 1 The lower bound in condition (4) prevents the existence of accumulation points in the sequence $\{t_k\}_{k=1}^{\infty}$, and, hence, avoids the existence of Zeno behavior, which is typically undesired in practice. The upper bound in condition (4) guarantees that the time in between samples is bounded above, i.e., persistent sampling. Notice that, the pairs of parameters $(T_1^{(i)}, T_2^{(i)})$ are in general different for different values of i. This enables our model to capture asynchronous aperiodic sampling events. The first bound in (4) allows to consider a jump at the initial time, i.e., $t_0 = 0$.

Our goal is to provide sufficient conditions for exponential stability for the closed-loop system obtained by interconnecting controller (3) with plant (1). Then, by defining the error $\tilde{\eta} = x_p - \eta$ and by taking as a state $\bar{x} = (x_p, \tilde{\eta})$, the closed-loop system can be modeled via the following dynamical system with jumps:

$$\begin{cases}
\dot{\bar{x}}(t) = A_{cl}\bar{x}(t) + B_{cl}\mathfrak{d}_{\bar{u}}\left(K_{cl}\bar{x}(t)\right) \ t \notin \mathcal{M} \\
\bar{x}(t^{+}) = \begin{bmatrix} x_{p}(t) \\ \tilde{\pi}(\bar{x}(t), t) \end{bmatrix} \qquad t \in \mathcal{M}
\end{cases}$$
(5)

where for all $t \in \mathcal{M}, \bar{x} \in \mathbb{R}^{2n}, \tilde{\pi}(\bar{x}, t) := x_p - \pi(x_p, x_p - \tilde{\eta}, t)$ and

$$A_{cl} \coloneqq \begin{bmatrix} A + BK & -BK \\ 0 & A \end{bmatrix}, K_{cl} \coloneqq \begin{bmatrix} K & -K \end{bmatrix}, B_{cl} \coloneqq \begin{bmatrix} B \\ 0 \end{bmatrix}$$

and $y \mapsto \mathfrak{d}_{\bar{u}}(y) \coloneqq \mathfrak{sat}_{\bar{u}}(y) - y$ is the decentralized deadzone nonlinearity. Notice that the definition of $\tilde{\pi}$ ensures that for all $(x_p, \tilde{\eta}, t) \in \mathbb{R}^{2n} \times \mathcal{M}$:

$$\tilde{\pi}_i(x_p^{(i)}, \tilde{\eta}^{(i)}, t) = \begin{cases} 0 & \text{if } t \in M_i \\ \tilde{\eta}^{(i)} & \text{otherwise} \end{cases}, \quad \forall i \in \mathcal{N}.$$

2.2 Hybrid modeling

Since the closed-loop system (5) experiences jumps when a new measurement is available, we rely on the hybrid dynamical model framework in [18]. In particular, we provide a single hybrid model capturing the behavior of the closed loop due to each possible evolution generated by any sequence $\{t_k^{(i)}\}_{k=1}^{\infty}$, for $i \in \mathcal{N}$, satisfying (4). This approach leads to a hybrid system with nonunique solutions allowing one to establish a result for all family of sequences $\{t_k^{(i)}\}_{k=1}^{\infty}$, for $i \in \mathcal{N}$, satisfying (4). The proposed modeling approach requires to model the hidden time-driven mechanism triggering the jumps of the controller. To this end, following [6], we add q timer variables $\tau_1, \tau_2, \dots, \tau_q$ to keep track of the duration of flows and to trigger jumps according to the mechanism in (5). To accomplish that, we make $\tau_1, \tau_2, \ldots, \tau_q$ decrease as ordinary time t increases and, whenever $\tau_i = 0$ for some $i \in \mathcal{N}$, reset it to any point in $[T_1^{(i)}, T_2^{(i)}]$, so to enforce (4). In particular, define $\tau := (\tau_1, \tau_2, \dots, \tau_q)$, $\mathcal{T} \coloneqq { \bigvee}_{i=1}^q [0, T_2^{(i)}], \, \text{and for all } i \in \mathcal{N}, \tau \in \mathcal{T}$

$$\widehat{\mathcal{G}}_{i}(\tau) := \left\{ \chi \in \mathbb{R}^{q} : \chi_{i} \in [T_{1}^{(i)}, T_{2}^{(i)}], \\ \chi_{k} = \tau_{k} \quad \forall k \in \mathcal{N} \setminus \{i\} \right\}.$$

$$(6)$$

Then, we consider the following dynamics for τ :

$$\dot{\tau} = -\mathbf{1}_q \qquad \tau \in \mathcal{T}
\tau^+ \in \bigcup_{i \in \{k \in \mathcal{N}: \ \tau \in \mathcal{D}_k\}} \widehat{\mathcal{G}}_i(\tau) \quad \tau \in \widehat{\mathcal{D}}$$
(7)

where one gets $\widehat{\mathcal{D}} := \bigcup_{i \in \mathcal{N}} \mathcal{D}_i$ with, for each $i \in \mathcal{N}$, $\mathcal{D}_i := \{\tau \in \mathcal{T} : \tau_i = 0\}$. In particular, the dynamics defined in (7) enable the timer variable to flow when $\tau \in \mathcal{T}$ and enforce a jump whenever, for some $i = 1, 2, \ldots, q$, $\tau_i = 0$. Specifically, at each jump only one component of the vector τ experiences a jump. Indeed, notice that when $\tau \in \mathcal{D}_i \cap \mathcal{D}_j$ for some $i \neq j \in \mathcal{N}$, only one component of the vector τ undergoes a change at the jump and multiple consecutive jumps occur. This phenomenon directly follows from having defined an outer semicontinuous jump map 1 .

We take as state $x := (\bar{x}, \tau) \in \mathbb{R}^{2n} \times \mathcal{T}$, which allows to consider the state \bar{x} in (5) along with the timer variable τ in (7). Define the flow map and the flow set as

$$f(x) := \begin{bmatrix} A_{cl}\bar{x} + B_{cl}\mathfrak{d}_{\bar{u}}(K_{cl}\bar{x}) \\ -\mathbf{1}_q \end{bmatrix}, \quad \forall x \in \mathcal{C} := \mathbb{R}^{2n} \times \mathcal{T}.$$
(8)

 $^{^{1}}$ A set-valued map F is outer semicontinuous if its graph is closed; see [18, Definition 5.9, Lemma 5.10].

To define the jump map, as a first step, define for all $i \in \mathcal{N}, \tilde{\eta} \in \mathbb{R}^n, h_i(\tilde{\eta}) \coloneqq \left(\bigoplus_{j=1}^q H_j^{(i)}\right) \tilde{\eta}$, where for each $i, j \in \mathcal{N}, H_j^{(i)} \coloneqq \begin{cases} \mathbf{0}_{n_j} \times n_j & \text{if } i = j \\ \mathbf{I}_{n_j} & \text{otherwise} \end{cases}$. Finally, from (6) and (7), one can define the jump map and the jump set as follows:

$$G(x) := \begin{bmatrix} x_p \\ \mathcal{J}(\tilde{n}, \tau) \end{bmatrix}, \quad \forall x \in \mathcal{D} := \mathbb{R}^{2n} \times \widehat{\mathcal{D}}$$
 (9)

with
$$\mathcal{J}(\tilde{\eta}, \tau) := \bigcup_{i \in \{k \in \mathcal{N}: \ \tau \in \mathcal{D}_k\}} \left(\{h_i(\tilde{\eta})\} \times \widehat{\mathcal{G}}_i(\tau) \right)$$
. The

jump map in (9) enables to capture the dynamics of the closed-loop system at the sampling times. Indeed, for all $x \in \mathcal{D}$, one has that $\tilde{\eta}_i^+ = 0$ if $\tau_i = 0$, and $\tilde{\eta}_i^+ = \tilde{\eta}_i$ otherwise.

Therefore, from (8) and (9), the closed-loop system reads:

$$\begin{cases} \dot{x} = f(x) & x \in \mathcal{C} \\ x^+ \in G(x) & x \in \mathcal{D}. \end{cases}$$
 (10)

The following lemma, whose proof is reported in [12], shows that (10) is well-posed ² thereby ensuring robustness with respect to small perturbations; see [18] for more details.

Lemma 1 Consider system (10). The sets C and D are closed, f is continuous, and G is locally bounded and outer semicontinuous on D. Consequently, the closed-loop system (10) is well-posed.

Existence of solutions to system (10) follows directly from the definition of the system, by checking that for every initial condition $\xi \in \mathcal{C} \cup \mathcal{D}$ there exists a nontrivial solution and every maximal solution to (10) from ξ is complete. Moreover, the definition of (10), similarly as in [9], ensures that for each solution, at most q jumps can occur consecutively without flowing. Such a property, along with (4), ensures that for every maximal solution ϕ to (10) and each $(t,j) \in \text{dom } \phi$ such that $(t,s) \in \text{dom } \phi$ for some $s \in \{j+1,j+2,\ldots,j+q\}$, one has $\left([t,t+\min\{T_1^{(1)},T_1^{(2)},\ldots,T_1^{(q)}\}]\times\{s\}\right) \subset \text{dom } \phi$. Essentially, the domain of the solutions to (10) manifests an average dwell-time property, with dwell time $\tau_D = \min\{T_1^{(1)},T_1^{(2)},\ldots,T_1^{(q)}\}$ and offset $N_0 = q$, which implies for each $(t,j) \in \text{dom } \phi$

$$\tau_D j \le t + q \tau_D \tag{11}$$

see, e.g., [18, Example 2.15]. Such a property prevents from the existence of Zeno solutions.

The goal of this paper is to characterize exponential stability of the origin for the \bar{x} substate of the closed loop (10), uniformly in τ . The fact that τ evolves in the compact set \mathcal{T} simplifies this task and allows stating it as a suitable stability property for the compact set

$$\mathcal{A} := \{0\} \times \mathcal{T} \subset \mathbb{R}^{2n} \times \mathbb{R}^{q}_{>0} \tag{12}$$

To this end, we adopt the notion of exponential stability of closed sets for hybrid systems as introduced in [38] and recalled next.

Definition 1 (Exponential stability [38]) The set \mathcal{A} is said to be exponentially stable for (10) if there exist positive real numbers κ, λ , and μ such that every maximal solution ϕ to (10) with $|\phi(0,0)|_{\mathcal{A}} \leq \mu$ is complete and

$$|\phi(t,j)|_{\mathcal{A}} \le \kappa e^{-\lambda(t+j)} |\phi(0,0)|_{\mathcal{A}} \qquad \forall (t,j) \in \operatorname{dom} \phi.$$
(13)

The set of points, denoted by $\mathcal{B}(\mathcal{A})$, such that each maximal solution to (10) from $\mathcal{B}(\mathcal{A})$ is complete and converges to \mathcal{A} , is denoted as the basin of attraction of \mathcal{A} . \diamondsuit

It is important to note that due to the presence of input saturation, the implicit objective is to characterize the basin of attraction of the closed set for which the exponential stability is guaranteed. Indeed, the basin of attraction can be $\mathbb{R}^{2n} \times \mathcal{T}$ only if the open-loop system, namely matrix A has some particular stability property; see [36] for these questions. The problem we solve is formalized as follows.

Problem 1 (Controller Design) Given matrices A and B of appropriate dimensions, and positive real numbers $T_1^{(1)}, T_2^{(1)}, \ldots, T_1^{(q)}, T_2^{(q)}$, design $K \in \mathbb{R}^{m \times n}$ such that the set A defined in (12) is exponentially stable for the hybrid system (10) and provide an inner approximation of the basin of attraction of A.

3 Lyapunov-Based Stability Analysis

To address Problem 1, we use the generalized sector condition for the deadzone nonlinearity presented in [37], which is recalled next in a convenient form.

Lemma 2 Let $X \in \mathbb{D}_+^m$, $L \in \mathbb{R}^{m \times 2n}$, and $z \in \mathbb{R}^{2n}$ be such that for all $i \in \{1, 2, ..., m\}$, $|L_{(i)}z| \leq \bar{u}_i$, where $\bar{u}_1, \bar{u}_2, ..., \bar{u}_m$ are defined in Section 2. Then, for each $u \in \mathbb{R}^m$:

$$\mathfrak{d}_{\bar{u}}(u)^{\mathsf{T}}X(\mathfrak{d}_{\bar{u}}(u) + u + Lz) \le 0. \tag{14}$$

 $^{^2\,}$ A hybrid system, like (10), is said to be well-posed if its set of solutions has good structural properties; see [18, Definition 6.27]. In simple words, well-posed hybrid systems are such that, in particular, every sequence of bounded solutions with converging initial conditions converge to a solution in the graphical sense.

To exploit the above result in our setting, it is convenient to state the following preliminary result.

Lemma 3 Let $\bar{\mu}, \sigma_1, \sigma_2, \ldots, \sigma_q \in \mathbb{R}_{>0}, R_1 \in \mathbb{S}_+^{n_1}, R_2 \in \mathbb{S}_+^{n_2}, \ldots, R_q \in \mathbb{S}_+^{n_q}, W \in \mathbb{S}_+^n, Z, J \in \mathbb{R}^{m \times n}$. Define:

$$\mathcal{L} := \left\{ \bar{x} \in \mathbb{R}^{2n} : |L_{(i)}\bar{x}| \leq \bar{u}_i, i \in \{1, 2, \dots, m\} \right\}$$

$$\mathcal{Q}_{\bar{\mu}} := \left\{ x = (\bar{x}, \tau) \in \mathcal{C} : \bar{x}^\mathsf{T} \hat{P}(\tau) \bar{x} \leq \bar{\mu} \right\}$$
(15a)

where for all $\tau \in \mathcal{T}$

$$L := \left[ZW^{-1} \ J \right] \qquad \widehat{P}(\tau) := W^{-1} \oplus R(\tau)$$

$$R(\tau) := \bigoplus_{i=1}^{q} R_i e^{\sigma_i \tau_i}$$
(15b)

and assume that

$$\begin{bmatrix} W & \mathbf{0} & Z_{(i)}^{\mathsf{T}} \\ \star & \bigoplus_{j=1}^{q} R_{j} & J_{(i)}^{\mathsf{T}} \\ \star & \star & \frac{\bar{u}_{i}^{2}}{\bar{x}} \end{bmatrix} \succeq 0 \qquad \forall i \in \{1, 2, \dots, m\}. \quad (16)$$

Then, the following inclusion holds:

$$Q_{\bar{\mu}} \subset \mathcal{L} \times \mathcal{T}. \tag{17}$$

PROOF. Define the following sets

$$\mathcal{R} := \left\{ \bar{x} \in \mathbb{R}^{2n} \colon \bar{x}^{\mathsf{T}} L_{(i)}^{\mathsf{T}} L_{(i)} \bar{x} \leq \bar{u}_{i}^{2}, i \in \{1, 2, \dots, m\} \right\}$$
$$\mathcal{E}_{\bar{\mu}}^{0} := \left\{ \bar{x} \in \mathbb{R}^{2n} \colon \bar{x}^{\mathsf{T}} \hat{P}(0) \bar{x} \leq \bar{\mu} \right\}$$

$$(18)$$

and observe that $\mathcal{R} \subset \mathcal{L}$. Let $\mathbb{J} := W^{-1} \bigoplus \mathbf{I} \bigoplus \mathbf{I}$, then by pre-and-post multiplying the left-hand side of (16) by \mathbb{J} one gets

$$\begin{bmatrix} W^{-1} & \mathbf{0} & W^{-1}Z_{(i)}^{\mathsf{T}} \\ \star & \bigoplus_{j=1}^{q} R_{j} & J_{(i)}^{\mathsf{T}} \\ \star & \star & \frac{\bar{u}_{i}^{2}}{\bar{\mu}} \end{bmatrix} \succeq 0 \qquad \forall i \in \{1, 2, \dots, m\}.$$

By Schur complement, (19) yields, for all $i \in \{1, 2, ..., m\}$ $\frac{1}{\bar{\mu}} \widehat{P}(0) - \frac{1}{\bar{u}_i^2} L_{(i)}^{\mathsf{T}} L_{(i)} \succeq 0$, that is $\mathcal{E}_{\bar{\mu}}^0 \subset \mathcal{R}$, which in turn implies

$$\mathcal{E}_{\bar{n}}^0 \subset \mathcal{L}.$$
 (20)

Now observe that for each $\tau \in \mathcal{T}$, one has $\widehat{P}(\tau) \succeq \widehat{P}(0)$. Therefore $\mathcal{Q}_{\bar{\mu}} \subset \mathcal{E}^0_{\bar{\mu}} \times \mathcal{T}$ which, thanks to (20), implies $\mathcal{Q}_{\bar{\mu}} \subset \mathcal{L} \times \mathcal{T}$. This concludes the proof.

Remark 2 Lemma 3 enables to exploit the sector condition in Lemma 2, which is typically coupled with quadratic Lyapunov functions, with "clock-dependent" Lyapunov functions. This key aspect of the paper clearly emerges in Theorem 1.

Now we are in a position to state the main result, which provides sufficient conditions for exponential stability of the closed-loop system in (10) together with an explicit estimate of the basin of attraction of \mathcal{A} .

Theorem 1 (Exponential stability) Let $K \in \mathbb{R}^{m \times n}$ be given. Suppose that there exist $W \in \mathbb{S}^n_+$, $R_1 \in \mathbb{S}^{n_1}_+$, $R_2 \in \mathbb{S}^{n_2}_+$, ..., $R_q \in \mathbb{S}^{n_q}_+$, $S \in \mathbb{D}^m_+$, $\sigma_1, \sigma_2, \ldots, \sigma_q, \bar{\mu} \in \mathbb{R}_{>0}$, $Z, J \in \mathbb{R}^{m \times n}$ such that (16) holds and that for all $\tau \in \mathcal{T}$

$$\begin{bmatrix}
\operatorname{He}((A+BK)W) & -BK & BS - WK^{\mathsf{T}} - Z^{\mathsf{T}} \\
\star & \operatorname{He}(R(\tau)A) - \Sigma R(\tau) & K^{\mathsf{T}} - J^{\mathsf{T}} \\
\star & \star & -2S
\end{bmatrix} \prec 0,$$

$$\underbrace{\mathbb{M}(\tau)}$$
(21)

where, for all $\tau \in \mathbb{R}^q$, $R(\tau) := \bigoplus_{j=1}^q e^{\sigma_j \tau_j} R_j$ and $\Sigma :=$

 $\bigoplus_{j=1}^{q} \sigma_{j} \mathbf{I}_{n_{j}}$. Then, the set \mathcal{A} defined in (12) is exponen-

tially stable for (10) and the set $Q_{\bar{\mu}}$ defined in (15a) is included in the basin of attraction of A. In particular, there exist positive numbers λ and κ such that for any maximal solution ϕ to (10), ϕ (0,0) $\in Q_{\bar{\mu}}$ implies

$$|\phi(t,j)|_{\mathcal{A}} \le \kappa e^{-\lambda(t+j)} |\phi(0,0)|_{\mathcal{A}} \quad \forall (t,j) \operatorname{dom} \phi.$$

PROOF. The proof of the result hinges upon [12, Theorem A.1] by showing that the hypotheses of the result imply all the conditions in [12, Theorem A.1].

Inspired by [10,13], for each $x \in \mathbb{R}^{2n+q}$ define

$$V(x) = x_p^{\mathsf{T}} W^{-1} x_p + \tilde{\eta}^{\mathsf{T}} R(\tau) \tilde{\eta} =: \bar{x}^{\mathsf{T}} \widehat{P}(\tau) \bar{x}$$
 (22)

where for all $\tau \in \mathbb{R}^q$, $R(\tau) := R_i e^{\sigma_i \tau_i}$. Notice that V is continuously differentiable on \mathbb{R}^{2n+q} and that for all $x \in \mathcal{C}$, $c_1|x|_A^2 \leq V(x) \leq c_2|x|_A^2$, with $c_2 := \max_{\tau \in \mathcal{T}} \lambda_{\max}(\widehat{P}(\tau))$ and $c_1 := \min_{\tau \in \mathcal{T}} \lambda_{\min}(\widehat{P}(\tau))$, which are both strictly positive due to P, R_1, \ldots, R_q being positive definite 3 . In addition, from [12, Lemma A.2] one has that for all $x \in \mathcal{D}, g \in G(x), V(g) - V(x) \leq 0$. Let \mathcal{L} be defined as in (15a). We now show that the satisfaction of (21)

³ By the definition of the system (10) and of the set \mathcal{A} , for every $x \in \mathcal{C} \cup \mathcal{D} \cup G(\mathcal{D})$, $|x|_{\mathcal{A}} = |(x_p, \tilde{\eta})|$.

implies that V is exponentially decreasing during flows. Pick any $\bar{x} \in \mathbb{R}^{2n}$ and $\tau \in \mathbb{R}^q$, then $\left\langle \frac{\partial}{\partial \tau} \tilde{\eta}^\mathsf{T} R(\tau) \tilde{\eta}, -\mathbf{1}_q \right\rangle = -\tilde{\eta}^\mathsf{T} \Sigma R(\tau) \tilde{\eta}$. Thus, for all $x \in \mathcal{C}$, one has

$$\begin{split} \langle \nabla V(x), f(x) \rangle &= x_p^\mathsf{T} \operatorname{He}(W^{-1}(A+BK)) x_p \\ &+ 2 x_p^\mathsf{T} W^{-1} B \mathfrak{d}_{\bar{\mathfrak{J}}_{\bar{u}}}(K_{cl}\bar{x}) \\ &+ \tilde{\eta}^\mathsf{T} \operatorname{He}(R(\tau)A) \tilde{\eta} \\ &- \tilde{\eta}^\mathsf{T} \Sigma R(\tau) \tilde{\eta} - 2 x_p^\mathsf{T} W^{-1} B K \tilde{\eta}. \end{split} \tag{23}$$

The proof is now completed by using an S-procedure argument; see, e.g., [3]. Specifically, pick any $x \in \mathcal{L} \times \mathcal{T}$. Then, from Lemma 2, the satisfaction of (16) implies that

$$\langle \nabla V(x), f(x) \rangle \le \langle \nabla V(x), f(x) \rangle - 2\Omega(\bar{x})$$
 (24)

where for all $\bar{x} \in \mathbb{R}^{2n}$

$$\Omega(\bar{x}) := \mathfrak{d}_{\bar{u}}(K_{cl}\bar{x})^{\mathsf{T}} S^{-1} (\mathfrak{d}_{\bar{u}}(K_{cl}\bar{x}) + K_{cl}\bar{x} + L\bar{x}) \tag{25}$$

and L is defined in (15b). Let for all $\bar{x} \in \mathbb{R}^{2n}$, $\varsigma(\bar{x}) := (\bar{x}, \mathfrak{d}_{3\bar{u}}(K_{cl}\bar{x}))$. Straightforward calculations show that for all $x \in \mathcal{C}$, the right-hand side of (24) can be rewrit-

ten as
$$\varsigma(\bar{x})^{\mathsf{T}}\underbrace{\begin{bmatrix} W^{-1} & 0 & 0 \\ \star & \mathbf{I} & 0 \\ \star & \star & S^{-1} \end{bmatrix} \mathfrak{M}(\tau) \begin{bmatrix} W^{-1} & 0 & 0 \\ \star & \mathbf{I} & 0 \\ \star & \star & S^{-1} \end{bmatrix}}_{\widehat{\mathfrak{M}}(\tau)} \varsigma(\bar{x})$$
 with

 $\widehat{\mathfrak{M}}(\tau)$ defined in (21). Namely, since (24) holds for any $x \in \mathcal{L} \times \mathcal{T}$, one has

$$\langle \nabla V(x), f(x) \rangle \le \varsigma(\bar{x})^{\mathsf{T}} \widehat{\mathfrak{M}}(\tau) \varsigma(\bar{x}) \quad \forall x \in \mathcal{L} \times \mathcal{T}.$$
 (26)

Pick $c_3 := -\max_{\tau \in \mathcal{T}} \lambda_{\max}(\widehat{\mathfrak{M}}(\tau))$, which is well defined due to the entries of $\widehat{\mathfrak{M}}$ being continuous functions of τ and strictly positive thanks to the satisfaction of (21). Then, using (26) one gets:

$$\langle \nabla V(x), f(x) \rangle \le -c_3 \bar{x}^\mathsf{T} \bar{x} \le -\frac{c_3}{c_2} V(x) \quad \forall x \in \mathcal{L} \times \mathcal{T},$$
(27)

where the last bound follows from the sandwich inequality shown here above and the definition of \mathcal{A} . Thus, V satisfies all the conditions in [12, Theorem A.1] with $\mathcal{U} = \mathcal{L}$. Therefore, the set \mathcal{A} defined in (12) is exponentially stable for (10) and each sublevel set of V included in $\mathcal{L} \times \mathcal{T}$ is contained in the basin of attraction of \mathcal{A} . To conclude the proof, observe that $\mathcal{Q}_{\bar{\mu}} = L_{V|_{\text{dom }V \cap \mathcal{C}}}(\bar{\mu})$ and that thanks to Lemma 3, (16) implies that $\mathcal{Q}_{\bar{\mu}} \subset \mathcal{L} \times \mathcal{T}$. This concludes the proof.

4 Control Design

This section devises a computationally tractable procedure for designing the controller gain K.

4.1 Computationally tractable sufficient conditions

Theorem 1 turns the stability analysis problem of the closed-loop system into the feasibility problem of a collection of matrix inequalities. Nonetheless, the search for feasible solutions to (21) is quite challenging since such conditions need to be checked over an uncountable set, i.e., \mathcal{T} . To overcome this problem, next we show that (21) can be turned without any additional conservatism into a finite collection of matrix inequalities. To this end, first consider the following preliminary result, whose proof is reported in [12].

Lemma 4 Let $\sigma_1, \sigma_2, \ldots, \sigma_q$ be given real positive numbers and for all $\tau \in \mathcal{T}$ define:

$$\Theta(\tau) := \bigoplus_{i=1}^{q} e^{\sigma_i \tau_i} \mathbf{I}_{n_i}. \tag{28}$$

Then, $rge \Theta$ is a polytope and in particular

$$\operatorname{rge}\Theta = \operatorname{co}\left\{\bigoplus_{i=1}^{q} \psi_{i} \mathbf{I}_{n_{i}} \colon \psi_{i} \in \{1, e^{\sigma_{i} T_{2}^{(i)}}\}.\right\}$$
(29)

Based on Lemma 4, the following proposition shows that (21) can be turned into a finite set of matrix inequalities.

Proposition 1 Let $W \in \mathbb{S}^n_+$, $K \in \mathbb{R}^{n \times m}$, $R_1 \in \mathbb{S}^{n_1}_+$, $R_2 \in \mathbb{S}^{n_2}_+$, ..., $R_q \in \mathbb{S}^{n_q}_+$, $S \in \mathbb{D}^m_+$, σ_1 , σ_2 , ..., $\sigma_q \in \mathbb{R}_{>0}$, $Z, J \in \mathbb{R}^{m \times n}$. Let for all $\tau \in \mathbb{R}^q$, $R(\tau) := \bigoplus_{j=1}^q e^{\sigma_j \tau_j} R_j$, $\Sigma := \bigoplus_{j=1}^q \sigma_j \mathbf{I}_{n_{p_j}}$, and $\widehat{R} := \bigoplus_{i=1}^q R_i$. Then, (21) holds if and only if for all $\Psi \in \mathcal{Z}$:

$$\mathfrak{N}(\Psi) := \begin{bmatrix} \operatorname{He}((A+BK)W) & -BK & BS - WK^{\mathsf{T}} - Z^{\mathsf{T}} \\ \star & \operatorname{He}(\widehat{R}\Psi A) - \Sigma \widehat{R}\Psi & K^{\mathsf{T}} - J^{\mathsf{T}} \\ \star & \star & -2S \end{bmatrix} \prec 0$$

(30a)

where:

$$\mathcal{Z} := \left\{ \bigoplus_{i=1}^{q} \psi_i \mathbf{I}_{n_i} \colon \psi_i \in \{1, e^{\sigma_i T_2^{(i)}}\} \right\}. \tag{30b}$$

PROOF. Since for all $\tau \in \mathcal{T}$, $R(\tau) = \widehat{R}\Theta(\tau)$ where $\tau \mapsto \Theta(\tau)$ is defined in (28), from (29) it follows that (21) holds if and only if $-\mathfrak{N}(\operatorname{rge}\Theta) \subset \mathbb{S}_{>0}$, where $\mathbb{S}_{>0}$ is the set of positive definite matrices of adequate dimensions. Thanks to Lemma 4, the latter is equivalent to require that:

$$-\mathfrak{N}\left(\operatorname{co}\left\{\bigoplus_{i=1}^{q}\psi_{i}\mathbf{I}_{n_{i}}:\psi_{i}\in\{1,e^{\sigma_{i}T_{2}^{(i)}}\}\right\}\right)\subset\mathbb{S}_{>0}. (31)$$

To conclude, it suffices to notice that since the application $\Psi \mapsto \mathfrak{N}(\Psi)$ is linear, from [12, Lemma A.2] it follows that

$$\mathfrak{N}\left(\operatorname{co}\left\{\bigoplus_{i=1}^{q}\psi_{i}\mathbf{I}_{n_{i}} \colon \psi_{i} \in \{1, e^{\sigma_{i}T_{2}^{(i)}}\}\right\}\right) = \operatorname{co}\mathfrak{N}\left(\left\{\bigoplus_{i=1}^{q}\psi_{i}\mathbf{I}_{n_{i}} \colon \psi_{i} \in \{1, e^{\sigma_{i}T_{2}^{(i)}}\}\right\}\right).$$

Thus, $S_{>0}$ being convex, the latter implies that (31) (or equivalently (21)) holds if and only (30a) holds. This concludes the proof.

The result given next enables to solve Problem 1 via the solution to some LMIs coupled to a line search on the parameters σ_i .

Proposition 2 Let
$$W \in \mathbb{S}_{+}^{n}$$
, $Y \in \mathbb{R}^{n \times m}$, $R_1 \in \mathbb{S}_{+}^{n_1}$, $R_2 \in \mathbb{S}_{+}^{n_2}$, ..., $R_q \in \mathbb{S}_{+}^{n_q}$, $S \in \mathbb{D}_{+}^{m}$, $\sigma_1, \sigma_2, \ldots, \sigma_q$, $\bar{\mu}, \alpha \in \mathbb{R}_{>0}$, and $Z \in \mathbb{R}^{m \times n}$. Define $\hat{R} := \bigoplus_{i=1}^{q} R_i$ and

$$\Sigma \coloneqq \bigoplus_{j=1}^{q} \sigma_{j} \mathbf{I}_{n_{p_{j}}}. If$$

$$\begin{bmatrix} W & \mathbf{0} & Z_{(i)}^{\mathsf{T}} \\ \star & \widehat{R} & 0 \\ \star & \star & \frac{\overline{u}_{i}^{2}}{\overline{\mu}} \end{bmatrix} \succeq 0 \qquad \forall i \in \{1, 2, \dots, m\}$$
 (32a)

and for all $\Psi \in \mathcal{Z}$, with \mathcal{Z} defined as in (30b)

$$\begin{bmatrix} \operatorname{He}(AW + BY) & -BY & BS - Y^{\mathsf{T}} - Z^{\mathsf{T}} & 0 \\ \star & -2\alpha W & Y^{\mathsf{T}} & \alpha \mathbf{I} \\ \star & \star & \star & -2S & 0 \\ \star & \star & \star & \operatorname{He}(\widehat{R}\Psi A - \Sigma \widehat{R}\Psi) \end{bmatrix} \prec 0.$$
(32b)

Then, $K = YW^{-1}$ solves Problem 1. In particular, the set \mathcal{A} defined in (12) is exponentially stable for (10) and the set $\mathcal{Q}_{\bar{\mu}}$ defined in (15a) is included in the basin of attraction of \mathcal{A} .

PROOF. As a first step, notice that (32a) corresponds to (16) with J=0. Moreover, following the same arguments as in the proof of Proposition 1 and using Y=KW, it follows that (32b) implies that for all $\tau \in \mathcal{T}$:

$$\begin{bmatrix} \operatorname{He}((A+BK)W) & -BKW & BS - WK^{\mathsf{T}} - Z^{\mathsf{T}} & 0 \\ \star & -2\alpha W & WK^{\mathsf{T}} & \alpha \mathbf{I} \\ \star & \star & -2S & 0 \\ \star & \star & \star & -Q(\tau) \end{bmatrix} \prec 0$$
(33)

where for all $\tau \in \mathcal{T}$, $Q(\tau) := -(\text{He}(R(\tau)A - \Sigma R(\tau)))$. Via Schur complement, the latter gives, for all $\tau \in \mathcal{T}$:

$$\begin{bmatrix} \operatorname{He}((A+BK)W) & -BKW & BS - WK^{\mathsf{T}} - Z^{\mathsf{T}} \\ \star & -2\alpha W + \alpha^{2}Q^{-1}(\tau) & WK^{\mathsf{T}} \\ \star & \star & -2S \end{bmatrix} \prec 0$$
(34)

Now observe that from (33), it follows that for all $\tau \in \mathcal{T}$, $Q(\tau) \succ 0$. Thus, by recalling the following elementary inequality

$$(\alpha Q(\tau)^{-1} - W)Q(\tau)(\alpha Q(\tau)^{-1} - W) \succeq 0 \quad \forall \tau \in \mathcal{T} \quad (35)$$

one has that for all $\tau \in \mathcal{T}$, $-WQ(\tau)W \leq \alpha^2 Q^{-1}(\tau) - 2\alpha W$. Therefore, using (34) one has, for all $\tau \in \mathcal{T}$

$$\begin{bmatrix} \operatorname{He}((A+BK)W) & -BKW & BS-WK^{\mathsf{T}}-Z^{\mathsf{T}} \\ \star & -WQ(\tau)W & WK^{\mathsf{T}} \\ \star & \star & -2S \end{bmatrix} \prec 0$$

Finally, by pre-and-post multiplying the matrix in the left-hand side of the above inequality by $\mathbf{I} \oplus W^{-1} \oplus \mathbf{I}$, one gets, for all $\tau \in \mathcal{T}$

$$\begin{bmatrix} \operatorname{He}((A+BK)W) & -BK & BS - WK^{\mathsf{T}} - Z^{\mathsf{T}} \\ \star & -Q(\tau) & K^{\mathsf{T}} \\ \star & \star & -2S \end{bmatrix} \prec 0$$

which reads as (21) with J = 0. Therefore, since (16) and (21) hold, by invoking Theorem 1 the result is proven.

Remark 3 Compared to Theorem 1, the estimate of the basin of attraction provided by Proposition 2 is in general more conservative. This is due to enforcing J=0 and using inequality (35). Therefore, once a control gain K is designed via Proposition 2, an additional analysis stage based on Proposition 1 can be performed to get a less conservative estimate of the basin of attraction. This aspect is discussed in Section 5.

4.2 Optimization Aspects

Similarly as in [22,36], with the objective of enlarging the basin of attraction of the set A, we turn our control design problem into the following optimization problem:

Problem 2 Design K to "maximize" the size of $Q_{\bar{\mu}}$. \diamond

To achieve this goal, one of the main difficulties that needs to be tackled is that the estimate of the basin of attraction provided by the set $Q_{\bar{\mu}}$ defined in (15a) is a subset of the set \mathcal{C} defined in (8). In other words, this estimate is intrinsically dependent on the "artificial" variable τ . This makes it difficult to define a convenient size

criterion to enlarge the basin of attraction in the " $(x_p, \tilde{\eta})$ -direction". To overcome this drawback, next we provide a subset of $Q_{\tilde{\mu}}$ for which a convenient size criterion can be formulated.

Lemma 5 Let $\bar{\mu} > 0$, $W \in \mathbb{S}_{+}^{n}$, $R_{1} \in \mathbb{S}_{+}^{n_{1}}$, $R_{2} \in \mathbb{S}_{+}^{n_{2}}$, ..., $R_{q} \in \mathbb{S}_{+}^{n_{q}}$, $S \in \mathbb{D}_{+}^{m}$, $\sigma_{1}, \sigma_{2}, \ldots, \sigma_{q}, \bar{\mu} \in \mathbb{R}_{>0}$ be given, $\mathcal{T} \ni \tau \mapsto \widehat{P}(\tau)$ be defined as in (15b), and $\tau^{\star} := (T_{2}^{(1)}, T_{2}^{(2)}, \ldots, T_{2}^{(q)})$. Then, $\mathcal{E}(\widehat{P}(\tau^{\star}), \bar{\mu}) \times \mathcal{T} \subset \{x \in \mathcal{C} : \bar{x}^{\mathsf{T}} \widehat{P}(\tau) \bar{x} \leq \bar{\mu}\}$.

PROOF. The proof follows directly from the fact that for each $\tau \in \mathcal{T}$, one has $\widehat{P}(\tau^*) \succeq \widehat{P}(\tau)$.

Leveraging Lemma 5, Problem 2 can be precisely formulated by selecting a suitable size criterion for the ellipsoidal set $\mathcal{E}(\widehat{P}(\tau^*), \bar{\mu})$. To this aim, different criteria are proposed in the literature; see [3]. In this work, we consider as a size criterion:

$$\operatorname{trace}\left(\frac{\widehat{P}(\tau^{\star})}{\bar{\mu}}\right) = \frac{\left(\operatorname{trace}(W^{-1}) + \sum_{i=1}^{q} \operatorname{trace}(R_i)e^{\sigma_i T_2^{(i)}}\right)}{\bar{\mu}}.$$
(36)

In particular, larger values of $\operatorname{trace}(\frac{1}{\bar{\mu}}\widehat{P}(\tau^{\star}))$ are associated to a smaller ellipsoid. To avoid dealing with W^{-1} as a decision variable, we consider the following additional constraint

$$\begin{bmatrix} M_W & \mathbf{I} \\ \star & W \end{bmatrix} \succeq 0 \tag{37}$$

which is equivalent to $M_W - W^{-1} \succeq 0$. Hence, by setting $\bar{\iota} = \bar{\mu}^{-1}$, one can replace (36) by:

$$\bar{\iota}\left(\operatorname{trace}(M_W) + \sum_{i=1}^q \operatorname{trace}(R_i)e^{\sigma_i T_2^{(i)}}\right).$$

Problem 2 can be solved by considering the following optimization scheme:

$$\begin{cases}
\min \operatorname{minimize} \varrho_{1}\bar{\iota} + \varrho_{2} \left(\operatorname{trace}(M_{W}) + \sum_{i=1}^{q} \operatorname{trace}(R_{i}) e^{\sigma_{i} T_{2}^{(i)}} \right) \\
\text{s.t.} \\
W \in \mathbb{S}_{+}^{n}, \bigoplus_{j=1}^{q} R_{j} \in \mathbb{S}_{+}^{n}, (30a), (32a), (37)
\end{cases}$$
(38)

where $\varrho_1, \varrho_2 \in \mathbb{R}_{>0}$ are tuning parameters.

When the real numbers $\alpha, \sigma_1, \sigma_2, \dots, \sigma_q$ are selected, optimization problem (38) is a semidefinite program, i.e., an optimization problem with linear objective and linear matrix inequality constraints. Therefore, the solution to

this problem can be efficiently obtained via numerical available software; see [3]. On the other hand, the nonlinearities involving the real numbers α , σ_i , $i = 1, 2, \ldots, q$ are easily manageable (at least when q is small enough) in a numerical scheme by performing a grid search.

Remark 4 For analysis purposes, an optimization problem wholly similar to (38) can be formulated using the conditions in Proposition 1 to get an estimate of the basin of attraction for a given control gain.

5 Numerical Example

In this section we showcase the effectiveness of the methodology proposed in this paper in an example ⁴. Consider the following data for system

(1):
$$A = \begin{bmatrix} -0.8 & -0.01 \\ 1 & 0.1 \end{bmatrix}$$
, $B = \begin{bmatrix} 0.4 \\ 0.1 \end{bmatrix}$, $\bar{u} = 1$, and

 $(T_2^{(1)}, T_2^{(2)}) = (0.3, 0.7)$. As a first step, we design the control gain K by solving optimization problem (38), with $\varrho_1 = \varrho_2 = 1$, via a line search on the parameters σ_1, σ_2 , and α . In particular, by selecting $\sigma_1 = 1.8$, $\sigma_2 = 2.3$, and $\alpha = 0.4$, one gets $K = \begin{bmatrix} -0.444 & -0.495 \end{bmatrix}$. As a second step, as suggested by Remarks 3 and 4, to reduce the conservatism, with the help of Lemma 4, we solve an optimization problem wholly similar to (38) based on the conditions in Theorem 1. Following this approach, by selecting $\sigma_1 = 3.8$ and $\sigma_2 = 2.3$, one gets

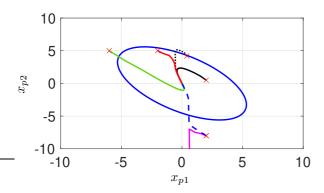
$$W^{-1} = \begin{bmatrix} 0.0983 & 0.0788 \\ 0.0788 & 0.0694 \end{bmatrix}, R_1 = 0.0141, R_2 = 0.01172,$$

and $\bar{\mu} = 2.66$. To better assess the conservatism in the estimation of the basin of attraction, one can analyze the following set of plant initial conditions:

$$\mathcal{B}_{x_p} := \left\{ x_p \in \mathbb{R}^n \colon x_p^\mathsf{T} N x_p \le \bar{\mu} \right\} \tag{39}$$

where $N := W^{-1} + (R_1 e^{\sigma_1 T_2^{(1)}} \oplus R_2 e^{\sigma_2 T_2^{(2)}})$. Notice that $\{(x_p, \tilde{\eta}) \in \mathbb{R}^{2n} \colon x_p = \tilde{\eta}, x_p \in \mathcal{B}_{x_p}\} \subset \mathcal{E}(P(\tau^*), \bar{\mu}),$ hence such a set is included in the basin of attraction of \mathcal{A} . Loosely speaking, the set \mathcal{B}_{x_p} contains all plant initial conditions such that when the controller is initialized to zero, the closed-loop state converges to \mathcal{A} . Fig. 1 depicts \mathcal{B}_{x_p} , along with some trajectories of the plant state from different initial conditions. In these simulations, jumps times are selected accordingly to a sinusoidal law with frequency 10. Fig. 1 points out several interesting aspects that are worthy of some comments. The first aspect is that $\mathcal{E}(P(\tau^*), \bar{\mu}) \times \mathcal{T}$ is in general not forward invariant for the closed-loop system as it

 $^{^4}$ Numerical solutions to LMIs are obtained through the solver SDPT3 [39] and coded in Matlab via YALMIP [25]. Simulations of hybrid systems are performed in Matlab via the $Hybrid\ Equations\ (HyEQ)\ Toolbox$ [33].



 \mathcal{B}_{xp} Set (blue ellipse) 1. and the response the plant state from different ini- $(0.4, 4.2, 0, 0, T_2^{(1)}, T_2^{(2)})$ tial conditions: (dash $(2.5, -2.5, 0, 0, T_2^{(1)}, T_2^{(2)})$ (solid black), black), $(2, -8, -1.8, -5.7, T_2^{(1)}, T_2^{(2)})$ (dashed-blue line).

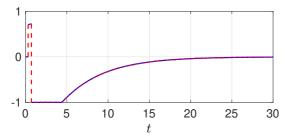


Fig. 2. Evolution of the control input from $x(0,0) = (-2,5,0,0,T_2^{(1)},T_2^{(2)}).$

is merely a subset of the (strongly) forward invariant set $\left\{x\in\mathcal{C}\colon\bar{x}^\mathsf{T}\widehat{P}(\tau)\bar{x}\leq\bar{\mu}\right\}$. This is testified by the trajectory from $x(0,0)=(0.4,4.2,0,0,T_2^{(1)},T_2^{(2)})$ (dashedblack line) that initially leaves $\mathcal{E}(P(\tau^\star),\bar{\mu})\times\mathcal{T}$. Another interesting aspect is illustrated by the trajectories from $x(0,0)=\xi_1:=(2,-8,0,0,T_2^{(1)},T_2^{(2)})$ (magenta line) and $x(0,0)=\xi_2:=(2,-8,-1.8,-5.7,T_2^{(1)},T_2^{(2)})$ (dashed-blue line). Although the two trajectories short belong to the estimate of the basin of attraction, i.e., $\mathcal{E}(P(\tau^\star),\bar{\mu})\times\mathcal{T}$, while ξ_2 does, despite the fact that $(2,-8)\notin\mathcal{B}_{x_p}$. This explains why the trajectory from ξ_2 converges to the origin, whereas the one from ξ_1 does not. In Fig. 2, we report the evolution of the control input from $x(0,0)=(-2,5,0,0,T_2^{(1)},T_2^{(2)})$. The figure emphasizes that, as foreseen through our methodology, despite control input saturation, as illustrated in Fig. 1 (red line), the plant state converges to zero.

6 Conclusion

Feedback control design for linear systems with asynchronous aperiodic sampled state measurements and actuator saturation was studied. The proposed controller contains a copy of the plant whose state is updated whenever a new measurement is available, hence performing a fusion of data obtained at different time instances. The approach we pursued relies on Lyapunov theory for hybrid systems and the use of quadratic clock-dependent Lyapunov functions. Sufficient conditions in the form of matrix inequalities are given to ensure regional exponential stability of a set containing the origin of the plant. Despite the difficulty inherent to the clock-dependence of the Lyapunov function used, a numerical procedure, based on semidefinite programming techniques allowed to design the control gain and to characterize an inner estimate of the basin of attraction of A in the form of ellipsoidal sets. Future research directions include the extension of the proposed methodology to the design of output feedback controllers in the presence of multirate output sampling and input saturation, for example by relying on the ideas presented in [10,26] and to the derivation of less conservative stability conditions.

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